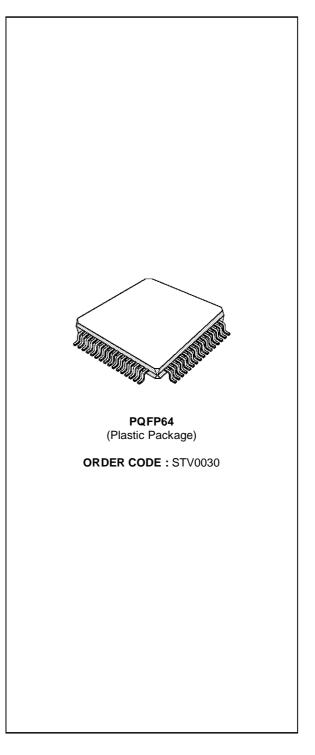


SATELLITE SOUND AND VIDEO PROCESSOR

- TWO INDEPENDENTLY PROGAMMABLE SOUND DEMODULATORS
- PLL DEMODULATION WITH 5-9MHz FRE-QUENCY SYNTHESIS
- 50/75µs, J17 OR NO DE-EMPHASIS PRO-GRAMMABLE OPTIONS
- DYNAMIC NOISE REDUCTION SYSTEM
- FIXED LEVEL AUXILIARY AUDIO INPUTS AND OUTPUTS
- VOLUME CONTROLLED AND MUTEABLE AUDIO OUTPUTS
- COMPOSITE VIDEO 6-BIT GAIN CONTROL
- COMPOSITE VIDEO SELECTABLE INVER-TER
- COMPOSITE VIDEO DEEMPHASIS AMPLI-FIER
- 9 x 5 VIDEO MATRIX AND SWITCHES WITH GRAPHICS INPUT
- 8-BIT DAC OUTPUT
- FULLY CONTROLLED VIA I²C BUS
- TWO HARDWARE SELECTABLE ADDRESSES (06_{HEX} and 46_{HEX})

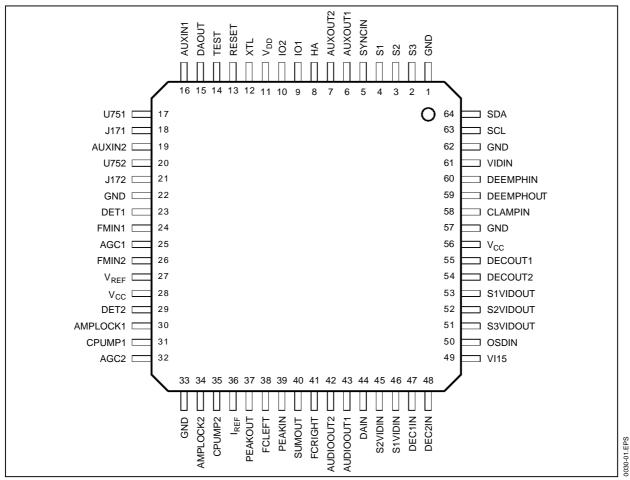


DESCRIPTION

The STV0030 is a BICMOS integrated circuit for use in satellite receivers. The great variety of FM deviations, energy dispersal and subcarrier standards makes the design of satellite receivers able to receive all programs a complex task. The device has been designed to specially adapt to all known conditions in analog video and radio transmissions.

The choice of the surface mounting PQFP64 package ensures optimisation of real-estate requirements.

PIN CONNECTIONS



PIN ASSIGNMENT

Pin	Name	Function	
1	GND	Digital Power Ground	
2	S3	SCART-3 Driver	
3	S2	SCART-2 Driver	
4	S1	SCART-1 Driver	
5	SYNCIN	Digital or Analog Sync Signal Input	
6	AUXOUT1	Fixed Level Audio Output Left	
7	AUXOUT2	Fixed Level Audio Output Right	
8	HA	Hardware Address	
9	IO1	Digital Input/Output 1	
10	IO2	Digital Input/Output 2	
11	V _{DD}	Digital 5V Power Supply	
12	XTL	4/8MHz Clock Input (optionally 4/8MHz Quartz Crystal)	
13	RESET	System Reset	
14	TEST	Test Pin	
15	DAOUT	DAC Amplifier Output	<u>ب</u>
16	AUXIN1	Auxilliary Audio Input Left	
17	U751	75µs De-emphasis Time Constant Left	0030-01.TBL



PIN ASSIGNMENT

Pin	Name	Function	
18	J171	J17 De-emphasis Time Constant Left	
19	AUXIN2	Auxiliary Audio Input Right	
20	U752	75µs De-emphasis Time Constant Right	
21	J172	J17 De-emphasis Time Constant Right	
22	GND	RF and Audio Ground	
23	DET1	FM PLL Filter Left	
24	FMIN1	FM Demodulator +ve Input	
25	AGC1	AGC Peak Detect Capacitor Left	
26	FMIN2	FM Demodulator -ve Input	
27	V _{REF}	2.44V Reference	
28	Vcc	Audio 12V Supply	
29	DET2	FM PLL Filter Right	
30	AMPLOCK1	Amplitude Detector Capacitor Left	
31	CPUMP1	FM PLL Charge Pump Capacitor Left	
32	AGC2	AGC Peak Detect Capacitor Right	
33	GND	Ground for Volume Control ANRS, VCO	
34	AMPLOCK2	Amplitude Detector Capacitor Right	
35	CPUMP2	FM PLL Charge Pump Capacitor Right	
36	I _{REF}	Current Reference Resistor	
37	PEAKOUT	ANRS Peak Detector Capacitor	
38	FCLEFT	Audio Roll-off Left	
39	PEAKIN	ANRS Peak Detector Input	
40	SUMOUT	ANRS Summing Output	
41	FCRIGHT	Audio Roll-off Right	
42	AUDIOOUT2	Level Controlled Audio Out Right	
43	AUDIOOUT1	Level Controlled Audio Out Left	
44	DAIN	Digital/Analog Converter Sense Input	
45	S2VIDIN	External Video Input 2	
46	S1VIDIN	External Video Input 1	
47	DEC1IN	Decoder 1 Input (e.g. D2MAC)	
48	DEC2IN	Decoder 2 Input (e.g. Videocrypt)	
49	VI5	Clamped Video Input	
50	OSDIN	On-screen Display Video Input	
51	S3VIDOUT	SCART-3 Video Output (with OSD)	
52	S2VIDOUT	SCART-2 Video Output (with OSD)	
53	S1VIDOUT	SCART-1 Video Output (without OSD)	
54	DECOUT2	Satellite Decoder Drive 2	
55	DECOUT1	Satellite Decoder Drive 1	
56	V _{CC}	Video 12V Supply	
57	GND	Video Ground	
58	CLAMPIN	Sync-tip Clamp Input	
59	DEEMPHOUT	Video De-emphasis Output	
60	DEEMPHIN	Video De-emphasis Input	
61	VIDIN	Video Input Buffer	
62	GND	Video Input Ground	
63	SCL	I ² C Bus Clock	
64	SDA	I ² C Bus Data	



INPUT/OUTPUT PIN CONNECTION

Pin N ^o	Circuitry	Comments
2 - 3 - 4	Pins 2 - 3 - 4 4 $50k\Omega$ $50V$ 4 $50k\Omega$ $50V$ 1_2 9 1_2 9 1_2 9 1_2 9 1_2	$\label{eq:linear_state} \begin{split} & \underline{As \ an \ input} \\ & I_1, \ I_2 \ off, \ Z_{IN} \geq 10 k \Omega \\ & \underline{As \ an \ output} \\ & V_{OUT} \leq 2V (L) I_1: \ OFF, \ I_2: \ ON \\ & V_{OUT} \geq 9.5V (H) I_1: \ ON, \ I_2: \ OFF \end{split}$
5	5 ^{-5kΩ} T T T T T T T T T T T T T	
6 - 7	Pins 6 - 7 100μΑ 9 700μΑ 9 700μΑ 9 700μΑ 9 7000μΑ	
8	000012 EPS	Input with TTL compatible levels
9 - 10	5V Pins 5V 9-10 F 9-10 F 9-10 F F F F F F F F F F F F F F F F F F F	Input/Output with TTL compatible levels
12		



4/30

INPUT/OUTPUT PIN CONNECTION (continued)

Pin N°	Circuitry	Comments
13	5V 10kΩ 13 ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓	V ₁₃ : L Reset active (I : ON)
14		To be grounded
15	5V 15 15	I _{15 (Max.)} = 2mA
16 - 19	Pins 16 - 19	To be externally DC biased to $V \cong V_{\text{REF}}$
17 - 18 20 - 21	Pins 17 - 18 20 - 21 I_1 I_2 I_2	$I_1 - I_2 = \frac{V_{AUD}}{18k}$
23 - 29	Pins 23 - 29 I_1	$I_1 - I_2 = f(\phi)$ ϕ : phase difference
24 - 26	24 CYOL VREF VREF VREF	Either Pin 24 or Pin 26 must be AC coupled to ground



Pin N^o Circuitry Comments I₁ 4μΑ to 'CA Pins 25 - 32 25- 32 30 - 34 $\mathsf{I}_1,\,\mathsf{I}_2$ current generator belong to the amplitude mixer Pins 30 - 34 150µA 10kΩ 0030-21.EPS V_{REF} 5V 27 27 $10 \mathrm{k}\Omega$ $V_{DC} = 2.44V$ 10kΩ 0030-22.EPS A 100μA \mathbf{I}_1 Reg 6 b4 : H, then I_1 , I_2 enabled. Pins 31 - 35 31 - 35 I₂ l₃, l₄ << l₁, l₂ 100μΑ 13 $\left| \right\rangle$ 0030-23.EPS $|I_4|$ 5V Recommanded value at Pin 36 : 240 k\Omega 36 0030-24.EPS (2.35V + V_{BE}) 36 37 DC voltage = f (audio amplitude) 5kΩ 37 0030-25.EPS 0030-11.TBL зv

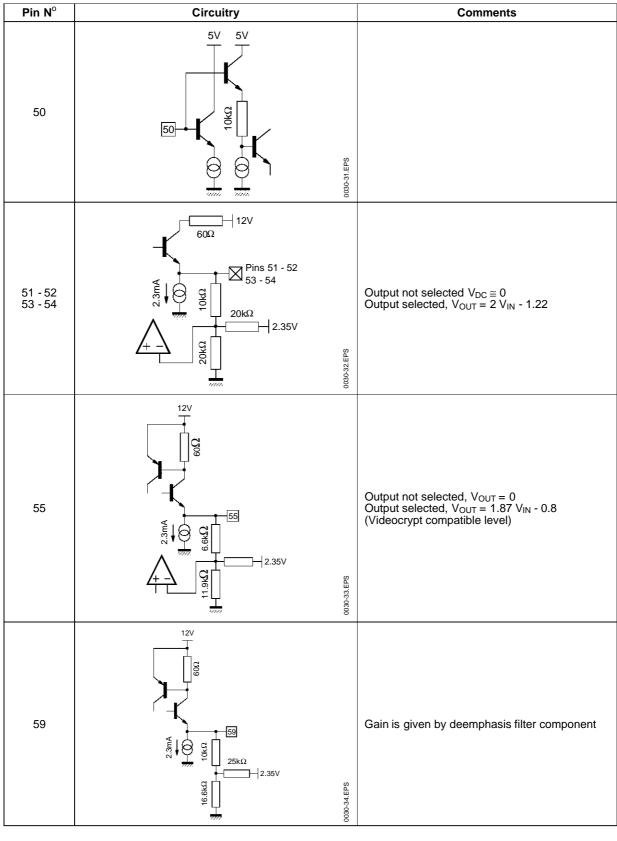
INPUT/OUTPUT PIN CONNECTION (continued)



Pin N ^o	Circuitry	Comments
38 - 41	Pins 38 - 41 OUT	I controlled by the amplitude of the audio signal ; I = $f(V_{37})$
40	V _{REF} (L) 99kΩ (R) (R) 60μA 99kΩ 60μA	$V_{OUT DC} = 2.44V$ V_{BE} needs to be AC coupled to the next stage
42 - 43	12V	V _{DC} = 4.7V Short circuit protected
44	DAC 44	V _{DC} = DAC output
15 - 46 - 47 48 - 49 - 58	Pins 15 - 46 - 47 48 - 49 - 58 $10k\Omega$	I_2 active if $V_{\text{PIN}} \leq 2.7 V$

INPUT/OUTPUT PIN CONNECTION (continued)

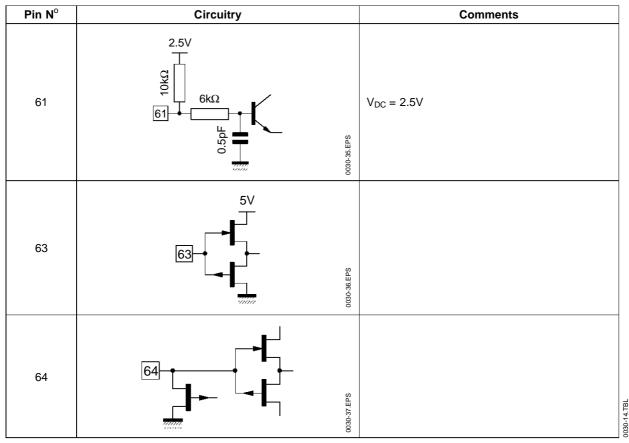




INPUT/OUTPUT PIN CONNECTION (continued)

SGS-THOMSON MIGROELECTRONICS 0030-13.TBL

8/30



INPUT/OUTPUT PIN CONNECTION (continued)

PIN FUNCTION DESCRIPTION

1 - SCART DRIVERS WITH OSD

'Text' must be able to be added to the video on command with or without 'blanking'. Blanking adds either a black line around the characters or a black box for the characters to sit in. During 'blanking' the chip will output OSD 'background' level.

There are 2 basic presentation modes of OSD 'text'.

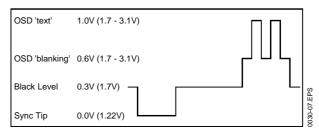
- OSD 'text' written on a blank screen of selectable brightness.
- 2) OSD 'text' written on top of the TV picture.

Additionally the OSD 'text' can be displayed with or without 'blanking'. The blanking from the microcontroller typically comes in two styles. Either as a line around characters or a box for the characters to sit in).

1 of 4 output levels from the SCART are required when using OSD. The diagram below shows part of a TV line displaying a small OSD. It starts with :

- a) a line sync pulse of 5μ s at 0V then,
- b) 13µs at black level, followed by

- c) 2µs at OSD 'background' level.
- d) This is the OSD 'text' which has been shown as 2µs pulses going to peak white. These would be typically programmed to say 0.8V with the OSD 'background' at 0.3V (1.7V at matrix output).



In order to achieve this two DACs are used to generate the programmable voltage levels for OSD 'background' and 'text'.

The video outputs and terms are :

- i) Normal video
 - sync tips = +1.22V black level = +1.7V peak white = +3.1V



ii) Video with OSD

Video with OSD sync tips = +1.22V black level = +1.7V 'background' level = +1.7V to 3.1V, programmable in 8 steps of 0.2V per step 'text' = +3.1V to 1.7V, programmable in 8 steps of 0.2V per step

- iii) 'background' occurs during active video so whenever the micro generates 'blanking' it will force the SCART output to 'background' level (+1.7V to +3.1V).
- iv) 'text' occurs during active video so whenever OSD occurs the SCART output is forced to 'peak white' (+3.1V to +1.6V). In fact the OSD 'text' has higher priority than 'background' so gets applied to the signal stream after 'background'.

Described above is an ideal situation with $2.0V_{PP}$ video with the bottom of the sync pulses at 1.22V. However the actual voltage of the sync tips will vary because the video is AC coupled and the picture content will change. Thus, in order to apply the 'blanking' or 'OSD' we need DC restored video the voltage of the sync tips are known and this is achieved by having 'DC level clamps' on all video input pins.

Pin 2 S3

This is the tri-stateable bi-directional SCART pin 8 driver for SCART-3. Since SCART-1,2 & 3 can be either controller or receiver SCART pin 8 is either an output or input. Since some equipment does not follow the SCART spec fully, the micro has full control of these pins via bits in the 'control' block.

As input; Z_{IN} >10k Ω <2V = logic 0, >9.5V = logic 1 As output; logic 0 outputs sink 1-3mA to gnd

logic 1 outputs source 1-3mA from V_{CC} -1V

Pin 3 S2

SCART-2 driver, same spec as pin 2.

Pin 4 S1

SCART-1 driver, same spec as pin 2.

2 - CONTROL BLOCK

This has an I^2C bus 2 wire interface with the related registers. It also contains 2 bi-directional tri-state-able logic I/O drives.

Pin 1 Ground

The main power ground connection for the control logic, registers, the I²C bus interface, synthesiser

& watchdog.

Pin 8 HA

For the l^2C bus interface it is necessary to have a programmable bit on the address in case the chip address clashes with that of another chip in the application. CMOS input levels must be held high or low externally; 0 = 06h, 1 = 46h.

Pin 9 IO1

A general purpose I/O pin.

Pin 10 IO2

A general purpose I/O pin. It may alternatively be programmed via I^2C bus to output an internally generated 20kHz squarewave for LNB control (ASTRA recommandation is 22kHz ± 4kHz).

Pin 11 V_{DD}

Digital +5V power supply.

Pin 63 SCL

This is the l^2 C bus clock line. Clock = DC to 100kHz. Requires external pull up eg. 10k Ω to 5V.

Pin 64 SDA

This is the I^2C bus data line. Requires external pull up eg. $10k\Omega$ to 5V.

3 - MISCELLANEOUS BITS

Pin 12 XTL

This pin allows for the on-chip oscillator to be either used with a crystal to ground of 4MHz, or to be driven by an external source via a $22k\Omega$ series resistor and 1nF series capacitor. This second option is recommanded. The external source can be either 4MHz or 8MHz. A programmable bit in the control block removes a÷ 2 block when the 4MHz option is selected.

Pin 13 RESET

Power on reset output. This is a voltage sensitive circuit and will give a LOW output until both supplies to this chip rise to greater than 80% of their correct value. It has some hysteresis so will remain HIGH until either of the supplies falls below 65%. The output is a current sink, 2mA, with an internal $10k\Omega$ pull-up resistor. It is possible to force the chip into or out of reset and by pass the on-chip delay.

Typical thresholds are:	V _{DD} =5V	Vcc=12V
power up	3.8V	9.2V
power down	3.2V	7.5V



10/30

Pin 14 TEST

Test pin to enable scan path testing of logic. Normal operation = L Test mode = H It must be grounded for normal operation.

Pin 15 DAOUT

This is the LNB drive which is the output of an amplifier that compares the input on pin 44 with an 8-bit DAC. The DAC is programmed through the I^2C bus interface to set the reference voltage for the LNB drive amplifier. The voltage range is 0 to 2.35 volts (2 bandgaps), hence the feedback signal must be divided down to be in this range. Pin 15 output current must be maintained lower than 2mA.

Pin 44 DAIN

The voltage actually fed to the LNB is divided down & connected here (feedback) for the on-chip regulator.

4 - SOUND DETECTION BLOCK

The different de-emphasis formats are selected through the audio MUX and are the same for both channels. That is, as with the volume control, the I^2C bus registers simultaneously control both channels with each bit of registers.

Pin 6 AUXOUT1 (LEFT)

This audio output is sourced directly from the audio MUX, and as a result does not include any volume control function. It will output $1V_{PP}$ biased at 2.35V with Z_{OUT} =1k Ω . If SCART socket drive is needed then an external buffer will be required.

Pin 7 AUXOUT2 (RIGHT)

See pin 6.

Pin 16 AUXIN1 (LEFT)

This pin allows an auxilliary audio signal to be connected to the input of channel 1 audio processor and hence makes use of the on chip volume control. An on-board MAC decoder is a typical user of this feature. Needs to be externally biased to V_{REF} .

Pin 17 U751 (LEFT)

A capacitor and resistor in parallel of 75 μ s time constant connected between here and V_{REF} (pin 27) to provide 75 μ s de-emphasis for channel 1. Internally selectable is an internal resistor that can be programmed to be added in parallel thereby converting the network to approx 50 μ s de-emphasis. The value of the internal resistor is $44k\Omega \pm 30\%$. The amplifier for this filter is voltage input, current output; with \pm 500mV input the output will be \pm 55 μ A.

Pin 18 J171 (LEFT)

The external J17 de-emphasis network for channel 1. The amplifier for this filter is voltage input, current output; with \pm 500mV input the output will be \pm 55µA.

Pin 19 AUXIN2 (RIGHT)

Same as Pin 16.

Pin 20 U752 (RIGHT) Same as Pin 17.

 Pin 21
 J172 (RIGHT)

 Same as Pin 18.

Pin 22 GND

This ground pin is double bonded ; 1) to channel 1 RF section & VCO and 2) to both AGC amplifiers, channel 1 audio section, audio MUX, internal power & references to audio section.

Pin 23 DET1 (LEFT)

The output of FM phase detector 1 (left channel). This is for the connection of an external loop filter for the PLL. The output is a push pull current source with \pm 90µA output with \pm 500mV input to the internal mixer.

Pin 24 FMIN1

This is the positive input to the two FM demodulators. It feeds two AGC amplifiers with a bandwidth of at least 5-10MHz. There is one amplifier for each channel both with the same differential input. The AGC amplifiers have a 0dB to + 40dB range ; in conjunction with AMPLOCK function it will produce a constant amplitude of the selected subcarrier. $Z_{IN} = 5k\Omega$ Min input = $2mV_{PP}$ per subcarrier Max input = $500mV_{PP}$ (max when all subcarriers are added together, when their phases coincide).

Pin 25 AGC1 (LEFT)

AGC amplifier 1 peak detector capacitor connection. The output current has an attack/decay ratio of 1:32. That is the ramp up current is approximately 5μ A and decay current is approximately 160μ A. 11V gives maximum gain. This pin is also driven by a circuit monitoring the voltage on AM-PLOCK1.



Pin 26 FMIN2

This is the -ve input to the AGC amplifiers forming a differential input stage. All inputs are biased to V_{REF} (2.35V) via an internal 10k Ω resistor.

Pin 27 VREF (2.35V)

This is the audio processor voltage reference used throughout the FM/audio section of the chip. As such it is essential that it is well decoupled to ground to reduce as far as possible the risk of crosstalk and noise injection. This voltage is derived directly from the bandgap reference. It cannot sink current.

Pin 28 V_{CC} (+12V power supply)

Double bonded main power pin for the audio/FM section of the chip. The two bond connections are; (1 to the ESD and guard rings and 2) to power the circuit and on chip regulators/references.

Pin 29 DET2 (RIGHT)

Same as Pin 23.

Pin 30 AMPLOCK1 (LEFT)

The output of amplitude detector 1. Requires a capacitor and a resistor to GND. The voltage across this is used to decide whether there is a signal being received by FM det 1. The level detector output drives a bit in the 'control block' I^2C bus register 7 bit 0. This also drives AGC amp 1. When the voltage on this pin is >(V_{REF}+1V_{BE}) it sinks current to V_{REF} from pin 25 to reduce the AGC gain.

Pin 31 CPUMP1 (LEFT)

The output from the frequency synthesizer is a push-pull current source which requires a capacitor to ground to derive a voltage to pull the VCO to the target frequency. The output is \pm 100µA to achieve lock and \pm 2µA during lock to provide a tracking time constant of approximately 10Hz.

Pin 32 AGC2 (RIGHT)

Same as Pin 25.

Pin 33 GND

This ground pin is double bonded; 1) to the volume control, ANRS, ESD and guard rings and 2) to the VCO & RF section of channel 2.

Pin 34 AMPLOCK2 (RIGHT)

Same as Pin 30

Pin 35 CPUMP2 (RIGHT)

Same as Pin 31.

Pin 36 IREF

This is a buffered V_{REF} output to an off-chip resistor to produce an accurate current reference, within the chip, for the biasing of amplifiers with current outputs into filters. It is also required for the ANRS circuit to provide accurate rolloff frequencies. This pin should not be decoupled as it will inject current noise. The target current is $10\mu A \pm 2$ thus a $240k\Omega \pm 1\%$ is required.

5 - AUTOMATIC NOISE REDUCTION SYSTEM (ANRS) AND VOLUME CONTROL

There is a ANRS defeat which allows the audio to bypass the ANRS section but still go to the volume control. The volume control will be 32 steps of 1.25dB, with the lowest step being audio 'mute' (no output).

Pin 37 PEAKOUT

The ANRS control loop peak detector output requires a capacitor to ground from this pin. Also a resistor to V_{REF} pin 27 to give some accurate decay time constant. The value of the capacitor and resistor control the attack and decay times, typically 0.1ms attack & 25ms decay.

Pin 38 FCLEFT

The variable bandwidth gm amplifier has a current output which is variable depending on the input signal amplitude as defined by the control loop. The output current is then dumped into an off-chip capacitor which together with the accurate current reference define the min/max rolloff frequencies. A capacitor is connected to ground from this pin for channel 1 (left).

Pin 39 PEAKIN

This is the input to the control loop peak detector and is connected to the output of the off-chip control loop band pass filter. This requires AC coupling between this input pin and the summer output Pin 40.

Pin 40 SUMOUT

The two audio demodulated signals are summed together with an amplifier with a gain of 0.5 ie. if both inputs are 1 volt then the output is 1 volt. This amplifier has an input follower buffer which gives a V_{BE} offset in the DC bias voltage. Thus the filter which this amplifier drives must include AC coupling to the next stage (pin 39).

Pin 41 FCRIGHT Same as Pin 38



12/30

Pin 42 AUDIOOUT2 (RIGHT)

The main audio output from the volume control level shifted and amplified to produce $2V_{PP}$ on a DC bias of 4.7 volts. This amplifier has short circuit protection and is intended to drive a SCART connector directly via AC coupling and meets the standard SCART drive requirements.

Pin 43 AUDIOOUT1 (LEFT)

Same as Pin 42.

6 - VIDEO BLOCK

Pin 5 SYNCIN

This input can be programmed to accept either a digital or an analog sync stream. The digital input will accept a made up sync signal consisting of the Genlock'line' and 'frame' sync output. If it is smaller than 0.8V then it produces a sync tip level voltage (1.22V). If it is greater than 2.0V it produces black level (1.52V). This will generate a sync stream signal consisting of syncs only, that will be available on the 'Video Matrix'. The DC level of these syncs at the SCART O/Ps will be sync tip = 1.22V black level = 1.7V. The analog path is connected directly to the video matrix.

When pin-49 is selected for the video MUX, it is still possible to insert an external sync. The input must be a digital sync (Reg 18b6 = 1 > sync enabled) and is typically used to stabilise the OSD on a bad or noisy video on pin-52 only.

Control bits

2 bits	I ² C bus Reg 1	5, bits 6		Effect
				Sync source
	default	Х	0	Digital
		0	1	Analog
		1	1	Pin-49 clamped

Pin 45 S2VIDIN

External video input $1V_{PP}$ AC coupled 75 Ω source impedance returned from a VCR for example. This input has a DC restoration clamp on its input. This gets the video to the correct DC level so that OSD can be added later when it is essential that the voltages of the video and OSD are at the correct levels with respect to each other. The clamp sink current is 1 μ A \pm 30% with the buffer Z_I > 1M Ω . This signal is an input to the Video Matrix and is called SCART-2 Return.

Pin 46 S1VIDIN

External video input $1.0 V_{PP}$ AC coupled 75Ω source impedance. This input has a DC restoration clamp on its input. This gets the video to the correct DC level so that OSD can be added later when it is

essential that the voltages of the video and OSD are at the correct levels with respect to each other. The clamp sink current is $1\mu A \pm 30\%$ with the buffer $Z_{IN} > 1M\Omega$. This signal is an input to the Video Matrix and is called SCART-1 Return.

Pin 47 DEC1IN

This input receives a PAL encoded output from, for instance a D2MAC decoder inside the receiver. The signals from pins 47 and 48 go to an internal selector switch that selects which signal goes to the Video Matrix.

Control bit

1 bit	I ² C bus Reg 1	bit	4	Signal selected
	default		0	Pin 48 (DEC2IN)
			1	Pin 47 (DEC1IN)

The output signal from the selector switch, which goes to the Video Matrix, is called Internal Decoder Return.

Pin 48 DEC2IN

This input can be driven for instance by the output of a Videocrypt decoder $ZOUT = 500\Omega$ video about $1.0V_{PP}$ from the reconstruction filter. This input has a DC restoration clamp on its input. This gets the video to the correct DC level so that OSD can be added later when it is essential that the voltages of the video and OSD are at the correct levels with respect to each other.

The clamp sink current is $1\mu A \pm 30\%$ with the buffer $Z_{IN} > 1M\Omega$. The signals from pins 47 and 48 go to an internal selector switch that selects which signal goes to the Video Matrix.

Control bit

1 bit	Reg 1	bit	4	Signal selected
	default		0	Pin 48 (DEC2IN)
			1	Pin 47 (DEC1IN)

Pin 49 VI5

A video input with clamp @ 1.22V internally. It may be employed for use with noisy video signals by overiding the sync clamp externally at about 3V (this overcomes the 1.22V plus $2x V_{BE}$).

Pin 50 OSDIN

Typically driven by a combination of an OSD O/P and the 'Blank' output from the OSD bit of the Micro. This input is then threshold detected and two outputs are generated.

Input	Output			
	Background	Text		
< 0.8V	Ō	0		
1.5-2.5V	1	0		
> 3.5V	1	1		



Pin 51 S3VIDOUT

Video driver for SCART-3 with OSD. This requires an external emitter follower buffer to drive a 150Ω load. The average DC voltage to be 1.5V on the O/P. The signal on pin 51 is video $2.0V_{PP}$ 5.5MHz B/W with sync tip=1.22V. This SCART O/P will be used to drive the TV typically. This pin gets its signal from the Video Matrix. It is then amplified and OSD is added before being output. The signal selected from the Video Matrix for output on this pin is controlled by a control register.

Control bits

3 bits I ² C bus Reg 4 bits	2,	1,	0	Source selected
default	_, 0 0 0 1 1	0 0 1 1 0 0	0 1 0 1 0 1	Baseband De-emhpasised Normal video Decoder return SCART 1 return SCART 2 return
	1	1	0 1	Syncs Nothing selected

OSD addition is controlled by a control register. It is possible to separately control whether or not to add 'blanking' and/or 'text' to the video stream to this SCART output.

Control bits

2

bits I ² C bus Reg 4 bits	4,	3	Effect enabled
default	0	0	Background = Off Text = Off
	0	1	Background = On Text = Off
	1	0	Background = Off Text = On
	1	1	Background = On Text = On

Pin 52 S2VIDOUT

(See 1/ SCART drivers with OSD). Video driver for SCART-2. See pin 51.

Control bits

3 bits I ² C bus Reg 3 bits	2,	1,	0	Source selected
default	0 0 0 1 1 1	0 0 1 1 0 0 1	0 1 0 1 0 1 0	Baseband De-emhpasised Normal video Decoder return SCART 1 return SCART 2 return Syncs Nothing selected
	L		1	and a second

OSD addition is controlled by a control register. It is possible to separately control whether or not to add 'blanking' and/or 'text' to the video stream to this SCART output.

Control bits 2 bits I²C bus Reg 3 bits 4, 3 Effect enabled

default		Δ	0	Background = Off Text = Off
uerault		0	0	Backyround = On Text = On
		0	1	Background = On Text = Off
		1	0	Background = Off Text = On
		1	1	Background = On Text = On
also pin-5 who	ere it is r	nossi	hle to a	add sync to stabilise OSD or vide

See also pin-5 where it is possible to add sync to stabilise OSD or video.

14/30



Pin 53 S1VIDOUT

Video driver for SCART 1. See pin 51. No OSD available on this output.

Control bits

3

bits I ² C bus Reg 2 bits	2,	1,	0	Source selected
default	0 0 0 1 1 1	0 0 1 1 0 0 1	0 1 0 1 0 1 0	Baseband De-emhpasised Normal video Decoder return SCART 1 return SCART 2 return Syncs Nothing selected
				5

Pin 54 DECOUT2

Drives a second decoder such as D2MAC and this is able to pass 10MHz. See pin 53.

Control bits

3 bits I^2C bus Reg 6 bits	2,	1,	0	Source selected
default	0 0 0 1 1 1	0 0 1 0 0 1	0 1 0 1 0 1 0	Baseband De-emhpasised Normal video Decoder return SCART 1 return SCART 2 return Syncs Nothing selected
	•	•		r toti ing selected

Pin 55 DECOUT1

This output can drive for instance a decoder. To allow for D2MAC it is able to pass 10MHz; ZOUT < 75Ω . Video on this pin will be 1.9V_{PP}. When a Videocrypt encoded signal is output to the decoder on this pin its sync tips must be at 1.46V_{DC}. This pin gets its signal from the Video Matrix. It is then level shifted and amplified by 1.9 before being output. The signal selected from the Video Matrix for output on this pin is controlled by a control register.

Control bits

3 bits I ² C bus Reg 5 bits	2,	1,	0	Source selected
	0	0	0	Baseband
	0	0	1	De-emhpasised
default	0	1	0	Normal video
	0	1	1	Decoder return
	1	0	0	SCART 1 return
	1	0	1	SCART 2 return
	1	1	0	Syncs
	1	1	1	Nothing selected



Pin 56 Vcc

+12V double bonded ; 1) ESD+guard rings & 2) video circuit power.

Pin 57 GND

Strategically placed video power ground connection to reduce video currents getting into rest of circuitry.

Pin 58 CLAMPIN

This pin clamps the most negative extreme of the input (the sync tips) to $+1.22V_{DC}$ (or appropriate voltage). The video at the clamp input is only $1V_{PP}$. This clamped video which is de-emphasised, filtered and clamped (energy dispersal removed) is normal, negative syncs, video. This signal drives the Video Matrix input called Normal Video.

It has a weak ($1.0\mu A \pm 15\%$) stable current source pulling the input towards GND. Otherwise the input impedance is very high at DC to 1kHz $Z_{IN} > 2M\Omega$. Video bandwidth through this is -1dB at 5.5MHz. The CLAMP input DC restore voltage is then used as a means for getting the correct DC voltage on

the SCART outputs.

Pin 59 DEEMPHOUT

Output of de-emphasis $Z_{OUT} < 50\Omega 2V_{PP}$ B/W still 10.25MHz. This is the output that drives the capacitor into the CLAMP. It has to do this via the sound removal filter which may be a 5 or 7 pole low pass filter $Z_{IN} = Z_{OUT} = 500\Omega$ -3dB corner about 5.25MHz. Video at pin 59 is positive. Internally the 2V_{PP} video is reduced to 1V_{PP} to drive the internal Video Matrix input called De-emphasised video. This signal also called 'Unclamped Unfiltered' and is the signal required by Filmnet PAL decoders for example. It is called unfiltered because it still has its high frequency sub-carriers (not been through the low pass filter).

Pin 60 DEEMPHIN

Input of de-emphasis stage, with a Z_{IN} of $10k\Omega$ or greater. B/W is 10.25MHz. The network between pins 59 & 60 will give 2x gain at 1.52MHz, about -2dB gain at 5MHz and +17dB gain at 10kHz as in accordance with CCIR 405-1.

Pin 61 VIDIN

AC-coupled Video input from a tuner. This is raw baseband up to 10.25MHz. Input amplitude is $0.25-1.0V_{PP}$ at 1.52MHz. $Z_{IN} > 5k\Omega$. This drives an on-chip video amplifier. The other input pin of this amp is AC grounded by being connected to an internal V_{REF} . The video amplifier has selectable gain from 0dB to 12.7dB in 63 steps. This is programmable, as is the output selected, whether normal or inverted. Control bits

default 0 0 0 0 0 0 0dB n n x 0.202dB	
1 1 1 1 1 1 +12.7dB	

The normal or inverted output selected (which is 1.0VPP) also drives the Video Matrix input called Baseband.

Control bit

1 bit I²C bus Reg 1 bit 3 Video selected

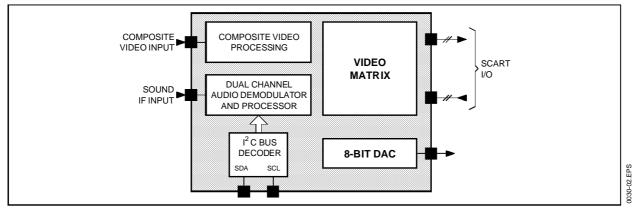
default ---- 0 Normal 1 Inverted

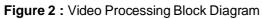
Pin 62 GND

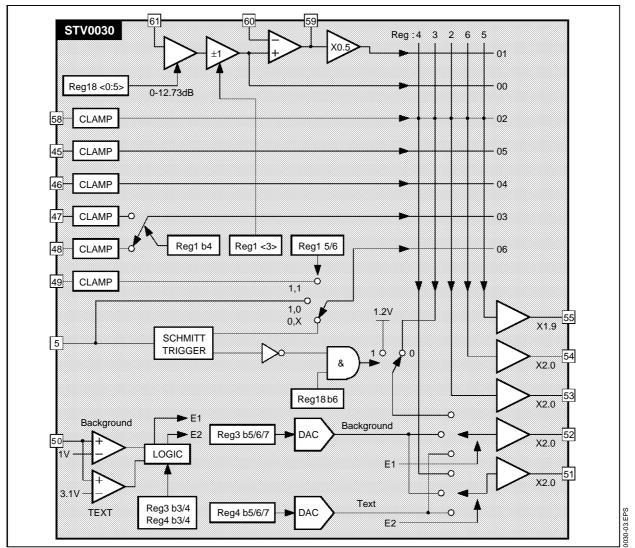
Ground, especially for the low level input from the tuner.



Figure 1 : Fundamental Block Diagram







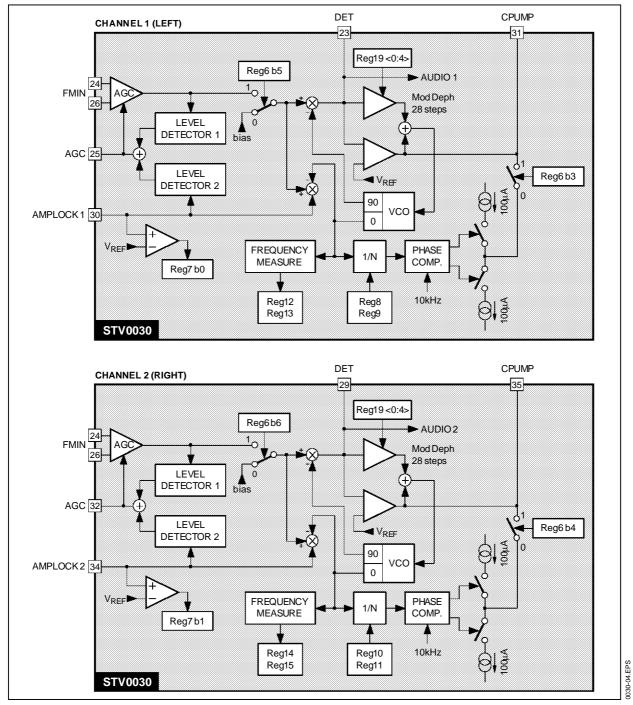
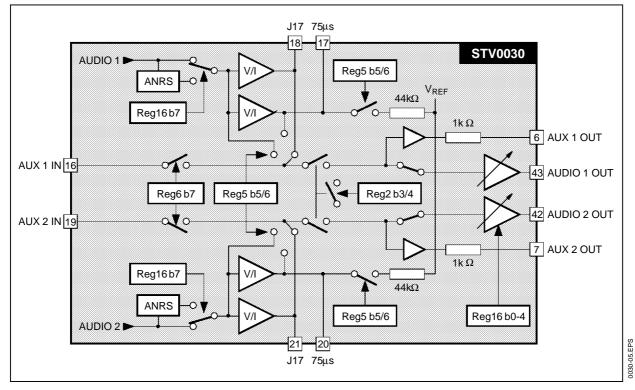


Figure 3 : FM Demodulation Block Diagram



Figure 4 : Audio Processing Block Diagram



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Units
V _{CC} V _{DD}	Supply Voltage	13.2 7.0	V V
P _{tot}	Total Power Dissipation	1.0	W
T _{oper}	Operating Temperature	0, +70	°C
T _{stg}	Storage Temperature	-55, +150	°C

THERMAL DATA

Symbol	Parameter	Value	Unit
R _{th(j-a)}	Thermal Resistance Junction-ambient	60	°C/W

DC AND AC ELECTRICAL CHARACTERISTICS

(V_{CC} = 12V, V_{DD} = 5V, $T_{amb} = 25^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V _{CC} V _{DD}	Supply Voltage Range		11.4 4.75	12 5.0	12.6 5.25	V V
IQ _{cc} IQ _{dd}	Supply Current			55 8	70 15	mA mA



DC AND AC ELECTRICAL CHARACTERISTICS (continued) $(V_{CC} = 12V, V_{DD} = 5V, T_{amb} = 25^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
AUDIO DE	MODULATOR (see Figure 3)					
FMI	FM Subcarrier Input Level Pin 24 - Pin 26 for AGC action	VCO locked on carrier at 6MHz, $680 k\Omega$ load on Pin 30 and Pin 34	5		500	mV _{PF}
DETH	Detector 1 and 2 (Pins 30 and 34) threshold for activating AGC			3.1		V
IDETA	Max sink/source Pins 30 and 34 current	With AGC operation		100		μΑ
IDETS	Max sink/source Pins 30 and 34 current	Without AGC limited gain		200		μΑ
VCOMI	VCO Mini Frequency	V _{CC} = 11.4 to 12.6V			5	MHz
VCOMA	VCO Maxi Frequency	$T_{amb} = 0 \text{ to } 70^{\circ} \text{C}$	10			MHz
AP50	1kHz Audio Level at PLL output Pins 23 and 29	$0.5V_{PP}$ 50kHz dev. FM input, Coarse deviation set to 50kHz (Reg 19 = 16 _{HEX})	0.5	1	1.5	V _{PP}
APA50	1kHz Audio Level at PLL output Pins 23 and 29	0.5V _{PP} 50kHz dev. FM input, Coarse and fine settings used	0.85	1	1.15	Vpp
DPCO	Digital Phase Comparator Pins 31 and 35 output current	Sink and source current to external capacitor		100		μA
UDIO DE	-EMPHASIS (1VPP on Detector 1 and 2 - P	ins 23, 29)				
D50e	1kHz Audio Level on 50µs de-emphasis output Pins 17 and 20	$22k\Omega$ and 2.2nF load Internal $44k\Omega$ off input and settings as APA50		1.91		Vpp
D50i	1kHz Audio Level on 50µs de-emphasis output Pins 17 and 20	$22k\Omega$ and 3.3nF load Internal $44k\Omega$ on input and settings as APA50		1.3		Vpp
D75	1kHz Audio Level on 50µs de-emphasis output Pins 17 and 20	$22k\Omega$ and 3.3nF load Internal $44k\Omega$ off input and settings as APA50		1.80		Vpp
DJ17	1kHz Audio Level on J17 de-emphasis output Pins 18 and 21	Load as in Application Diagram, input and settings as APA50		1.45		Vpp
R50	Internal resistor for 50/75µs switching		30	44	58	kΩ
UTOMAT	IC NOISE REDUCTION SYSTEM					
LRS	Left + Right Summer output Pin 40	1VPP on left and right channel	0.9	1	1.1	V _{PP}
NDRT	PIN 37 Level Detector Rise Time Constant	External 22nF load		110		μs
NDFT	PIN 37 Level Detector Fall Time Constant	External 22nF to GND and $1.2M\Omega$ to Pin 27		26.4		ms
NDLL	Pin 37 Bias Level	No audio in		2.44		V
NDML	Pin 37 Max. Level (internal clamp)	$3V_{\text{PP}}$ 1kHz on Pins 23 and 29	2.7	3	3.2	V
LLCF	Noise Reduction Cut-off Frequency for Low Level Audio	$100mV_{PP}$ on Pin 23 and 29		0.85		kHz
HLCF	Noise Reduction Cut-off Frequency at High Level	$1V_{\text{PP}}$ on Pin 23 and 29		7		kHz
	ITPUT (Pins 42 and 43)					
DCOL	DC Output Level Measured on Pins 42, 43	Aux. input Pins 16 and 19 connected to V_{REF}		4.7		V
AOLN	Audio Output Level at nominal input at 0dB attenuation setting, measured on Pins 42 and 43	FM input as for APA50 No de-emphasis No noise reduction	1.57	1.9	2.25	V _{PP}
AOL50	Audio Output Level at nominal input at 0dB attenuation setting, Measured on Pins 42 and 43	FM input as for APA50 50μs de-emphasis as for D50i No noise reduction	1.75	2.5	3.25	Vpp
AOL50B	Audio Output Level at nominal input at 0dB attenuation setting, Measured on Pins 42 and 43	FM input as for APA50 50μs de-emphasis as for D50i No noise reduction	1.55	2.5	3.45	V_{PP}

0030-06.TBL

DC AND AC ELECTRICAL CHARACTERISTICS (continued) $(V_{CC} = 12V, V_{DD} = 5V, T_{amb} = 25^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
AUDIO OUT	PUT (Pins 42 and 43) (continued)					
AOL17	Audio Output Level at nominal input at 0dB attenuation setting, Measured on Pins 42 and 43	FM input as for APA50, J17 de-emphasis No noise reduction	1.90	2.75	3.60	V _{PP}
AMA1	Audio Output Attenuation with mute on	1V _{PP} - 1kHz from Aux input Pins 16, 19 to Pins 42, 43		60		dB
MXAT	Max. Attenuation before mute	1kHz, from Aux input Pins 16 and 19 to Pins 42 and 43	30.7	32.7	34.7	dB
MXAG	Max. Audio Gain	1kHz, from Aux input Pins 16 and 19 to Pins 42 and 43		1.9		
ASTP	Attenuation of each of the 31 steps	1kHz		1.25		dB
THDA1	THD at 0dB Attenuation Setting Measured on Pins 42, 43	1V _{PP} - 1kHz from Aux. Input Pins 16, 19 to Pins 42, 43		0.15		%
THDA2	THD at 0dB Attenuation Setting Measured on Pins 42, 43	2V _{PP} - 1kHz from Aux. Input Pins 16, 19 to Pins 42, 43		0.3	1	%
THDFM	THD at 0dB Attenuation Setting Measured on Pins 42, 43	FM Input as for APA50 50µs de-emphasis, no ANRS		0.3	1	%
ACS	Audio Channel Separation Measured on Pins 42, 43	1V _{PP} - 1kHz from Aux. Input Pins 16, 19		80		dB
ACSFM	Audio Channel Separation at 1kHz - Measured on Pins 42, 43	 0.5V_{PP} 50kHz deviation FM input on one channel 		60		dB
		 0.5V_{PP} no deviation FM input on the other channel 				
		- Chip deviation set to 50kHz				
		- 50µs de-emphasis, no ANRS				
SNFM	Signal to Noise Ratio Measured on Pins 42, 43	FM input as for APA50 50µs de-emphasis, no ANRS weighted noise curve A		62		dB
SNFMNR	Signal to Noise Ratio Measured on Pins 42, 43	FM input as for APA50 No de-emphasis, ANRS ON weighted noise curve A		80		dB

AUXILIARY AUDIO OUTPUT (Pins 6 and 7)

DCOLAO	DC Output Level Measured on Pins 6, 7	Aux. Input Pins 16 and 19 connected to V_{REF}		2.35		V
AGAO	Audio Gain Measured on Pins 6, 7	1kHz from Aux. Inputs Pins 16, 19 to Pins 6, 7		1		
ZAO	Pins 6 and 7 Output Impedance		0.7	1	1.3	kΩ
THDA02	THD on Pins 6, 7	2V _{PP} - 1kHz from Aux. Input Pins 16, 19 to Pins 6, 7		0.01	0.1	%
THDAOFM	THD on Pins 6, 7	FM input as for APA50 50μs de-emphasis, no ANRS		0.15		%

I/O's (Pins 2, 3, 4, 9 and 10)

	•						
SCIL	Pin 2, 3 or 4 Low Level Input				2	V]
SCIH	Pin 2, 3, or 4 High Level Input		9.5			V	
SCOH	Pin 2, 3 or 4 High Level Output	10k Ω load to ground	9.5	11		V	
SCOL	Pin 2, 3 or 4 Low Level Output	$10k\Omega$ load to ground		0.1	2	V]
VIL	Pin 9 or 10 Low Level Input				0.8	V]
VIH	Pin 9 or 10 High Level Input		2.4			V]_
V _{OL}	Pin 9 or 10 Low Level Output	I _{Sink} = 2mA		0.2	0.4	V	14
V _{OH}	Pin 9 or 10 High Level Output	I _{Source} = 2mA	3.2	4.6		V	0500



DC AND AC ELECTRICAL CHARACTERISTICS (continued) $(V_{CC} = 12V, V_{DD} = 5V, T_{amb} = 25^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
RESET						
RSR	Reset Internal Pull-up Resistor	Reset high		10		kΩ
RSI	Reset Internal Sink Current	Reset low		2		mA
RTCCU	End of Reset for V _{CC}	$V_{DD} = 5V, V_{CC}$ going up		9.2		V
RTCCD	Start of Reset Threshold for V _{CC}	$V_{DD} = 5V, V_{CC}$ going down		7.5		V
RTDDU	End of Reset Threshold for VDD	V _{CC} = 12V, V _{DD} going up		3.8		V
RTDDD	Start of Reset Threshold for VDD	V _{CC} = 12V, V _{DD} going down		3.2		V
D/A CONV	ERTER (see Figure 1)					
DA00	Min. Output Voltage on Pin 15 for 00 _{HEX}	Sense input Pin 44 connected to output Pin 15, ILOAD <2mA	0	30	70	mV
DAFF	Max. Output Voltage on Pin 15 for FF _{HEX}	As for DA00	2.15	2.35	2.55	V
ILE	Integral Linearity			0.5	2	LSB
COMPOSI	TE SIGNAL PROCESSING (see Figure 2)					
VIDC	Pin 61 DC Level	External load current <1µA	2.15	2.35	2.55	V
ZVI	Pin 61 Input Impedance			10		kΩ
DEODC	Pin 59 DC Output Level	$1.2k\Omega$ from Pin 59 to Pin 60, $2k\Omega$ and $3k\Omega$ in series to $1\mu F$ from Pin 60 to GND	3.3	3.7	4.1	V
DEOAC	Pin 59 AC Level for GV = 0 dB	Pin 61 level = 1V _{PP} , 100kHz	1.87	2	2.13	Vpp
DEOMX	Pin 59 Max. AC Level before clipping	GV = 0 dB	5	6		VPP
DEOISC	Pin 59 Max. Source Current			8		mA
DEOISK	Pin 59 Max. Sink Current			2		mA
ZDEO	Pin 59 Output Impedance @ 5MHz			2		Ω
DGV	Gain Error vs GV @ 100kHz	For GV = 0 to 12.7dB	-0.5	0	0.5	dB
DINV	Gain Variation when using inverter	@ 100kHz	-0.5	0	0.5	dB
DEBW	Bandwidth at Pin 59 for 1V _{PP} input measured on Pin 59	@ - 3dB with GV = 0dB	10	20		MHz
DFG	Differential Gain on sync pulses measured on Pin 59	1V _{PP} CVBS + 0.5V _{PP} 25Hz sawtooth input Pin 61, GV=0			1	%
CLAMP ST	AGES					
ISKC	Clamp Input Sink Current Pins 45, 46, 47, 48 49 and 58	V _{IN} = 3V	0.6	1	1.4	μA
ISCC	Clamp Input Source Current same pins as ISKC	V _{IN} = 2V	30	50	70	μΑ
VCL	Sync Tip Level on selected output pin (Pins 51, 52, 53 or 54)	1V _{PP} CVBS through 10nF on input (45,46,47 or 58)	1.1	1.22	1.34	V
VCL5	Sync Tip Level on Pin 55	Same as for VCL	1.38	1.46	1.54	V
VIDEO MA	TRIX					
ХТК	Output Level on any output when 1V _{PP} CVBS input is selected for any other output	@ 5MHz		60		dB
BGT	Background Graphics Threshold Pin 50		0.8	1	1.5	V
CGT	Character Foreground Threshold Pin 50		2.5	3	3.5	V
BMN	Background Level for 000 selection	Measured on selected output Pin 52 or 53		1.7		V
BMX	Background Level for 111 selection	Same as BMN		3.1		V
FMN	Foreground Level for 000 selection	Same as BMN		1.7		V
FMX	Foreground Level for 111 selection	Same as BMN		3.1		V
BFG	Output buffer gain Pins 45,46,47,48,58	@ 100kHz	1.87	2	2.13	
BFG5	Output buffer gain Pin 55	@ 100kHz	1.77	1.9	2.03	



CIRCUIT DESCRIPTION : VIDEO SECTION

The composite video is first set to a standard level by means of a 64 step gain controlled amplifier. In the case that the modulation is negative, an inverter can be switched in. The deemphasis network is fed by a wide bandwidth amplifier and energy dispersal is removed by a sophisticated sync tip clamping circuit. This circuit is used on all inputs to a video switching matrix, thus making sure that no DC steps occur when switching video sources.

The matrix can be used to feed video to and from decoders such as D2MAC or scrambled analog video.

Two special inputs allow insertion of graphics on video or clean sync, or a mix of both even when only noise is present at the tuner output.

CIRCUIT DESCRIPTION : AUDIO SECTION

The two audio channels are totally independent except for the possibility given to output on both channels only one of the selected input audio channels.

To allow a very cost effective application, each channel uses PLL demodulation. Except for the overall high pass filter removing the video of the composite signal, no complex filter is needed.

The frequency of the demodulated subcarrier is chosen by a frequency synthesiser which sets the frequency of the internal local oscillator by comparing its phase with the internally generated reference. When the frequency is reached, the microprocessor switches in the STV0030 PLL and the demodulation starts. At any moment the microprocessor can read from the device the actual frequency to which the PLL is locked. It can also verify that a carrier is present at the given frequency, thanks to an amplitude demodulator which is also used for the audio input AGC.

In order to maintain constant amplitude of the recovered audio regardless of variations between satellites or subcarriers, the PLL loop gain may be programmed from 28 values, with coarse (four values) and fine (7 steps) adjustment.

Two different networks can be permanently connected for either 75µs or J17 de-emphasis. If 50µs de-emphasis is required, this can be inserted by an internal switch, thus allowing a worldwide application.

A dynamic noise reduction system (ANRS) is used using a lowpass filter, the cutoff frequency of which is controlled by the amplitude of the audio after insertion of a bandpass filter.

Two audio outputs are provided : one is a fixed $1V_{PP}$ and the other is a gain controlled $2V_{PP}$. These outputs are chosen by an audio matrix between non de-emphasised, 50 or 75µs de-emphasised, J17 deemphasisedaudio channel. In each case the dynamic noise reduction system can be used. Alternatively the selected outputs can receive the auxilliary audio inputs without processing. The gain controlled amplifier has a gain range from + 6dB to - 32dB with 1.25dB steps. This can also be muted.

The 8-bit Digital to Analog Converter outputs a voltage ranging from 0 and 2.35V via a differential amplifier for digital words from 00 to FF. By adding a suitable output buffer/converter, this DAC can be used to control LNB voltages (or polariser currents) with a range exceeding 0 to 2.35V.

IIC PROTOCOL

The chip address is 06_H if Pin 8 to ground and 46_H if Pin 8 to V_{DD}.

Transmit to Chip

······															
<u>S</u> =	S = START A = ACK P = STOP														
5	S CHIP ADDR W A REG ADDR A DATA A												() 	F	2
												/	/		
Rec	Receive from Chip														
S CHIP ADDR W A REG ADDR A POINTER A P S CHIP ADDR R A DATA A											A	Ρ			
	READ n BYTES														
In re	In read mode a pointer is written to register 0. This pointer addresses a particular register whose contents														

In read mode a pointer is written to register 0. This pointer addresses a particular register whose contents can be read any number of Times. There are 19 registers in Total.



IIC CONTROL REGISTERS

Reg 1 IIC \rightarrow reg

bit

- 0 L Not used
- 1 L Not used
- 2 L Not used
- 3 L Select video invert (H = inverted, L = non-inverted)4
 - (H = Pin 47, L = Pin 48)L Select input bits
- 5 L Select sync source
- 6 5 Х 0 digital sync input (Schmitt Trigger)
 - analogue sync input 1
- 0 Pin 49 clamp input 1 1
- L Select clamp Pin 49 6
- 7 Not used

Reg 2 IIC \rightarrow reg

bit

- Select source for SCART 1 O/P L 0
- Н Select source for SCART 1 O/P See Pin 53 description 1
- Select source for SCART 1 O/P 2 L
- 3 Н Select Left/Right/Stereo (see audio mux truth table)
- 4 н Select Left/Right/Stereo (see audio mux truth table)
- 5 Test multiplex control VCOCLK1/Div1000 L
- Test multiplex control VCOCLK2/Div1000 6 L
- 7 L Test multiplex control RESET

Reg 3 IIC \rightarrow reg

bit

- 0 L Select source for SCART 2 O/P
- H Select source for SCART 2 O/P See Pin 52 description 1
- 2 Select source for SCART 2 O/P L
- Select OSD effect for SCART 2 (Dackground) 3 L
- 4 Select OSD effect for SCART 2 (text) L
- 5 Select OSD "background" level (LSB) L
- Select OSD "background" level 6 L
- Select OSD "background" level (MSB) 7 L

Reg 4 IIC \rightarrow reg

bit

- 0 L Select source for SCART 3 O/P
- 1 H Select source for SCART 3 O/P See Pin 51 description
- 2 L Select source for SCART 3 O/P
- 3 L Select OSD effect for SCART 3 O/P (background)
- 4 L Select OSD effect for SCART 3 O/P (text)
- 5 H Select OSD "text" level (LSB)
- 6 H Select OSD "text" level
- 7 H Select OSD "text" level (MSB)



Reg 5 IIC \rightarrow reg

bit

- 0 L Select source for decoder 1 O/P
- 1 H Select source for decoder 1 O/P See Pin 55 description
- 2 L Select source for decoder 1 O/P
- 3 L Not used
- 4 L Not used
- 5 H Select de-emphasis (see audio mux truth table)
- 6 L Select de-emphasis (see audio mux truth table)
- 7 Not used

$\textbf{Reg 6} \quad \textbf{IIC} \rightarrow \textbf{reg}$

bit

- 0 L Select source for decoder 2 O/P
- 1 H Select source for decoder 2 O/P See Pin 54 description
- 2 L Select source for decoder 2 O/P
- 3 H Select frequency synth 1 OFF/ \overline{ON} (L=OFF)
- 4 H Select frequency synth 2 OFF/ON (L=OFF)
- 5 L Select RF source (L=OFF) to FM Det 1
- 6 L Select RF source (L=OFF) to FM Det 1
- 7 L Select aux audio input for both channels (see audio mux truth table)

$\textbf{Reg 7} \quad \textbf{Reg} \rightarrow \textbf{IIC}$

bit

- 0 Status of Amp_lock 1 (subcarrier presence)
- 1 Status of Amp_lock 2 (subcarrier presence)
- 2 Not used
- 3 Not used
- 4 Not used
- 5 Not used
- 6 Not used
- 7 Not used

Reg 8 IIC \rightarrow reg

bit

- 0 L Not used
- 1 L Not used
- 2 L Not used
- 3 L Select data direction for I/O 1 (H = output)
- 4 L Select data direction for I/O 2 (H = output)
- 5 L Not used
- 6 L Select frequency for det 1, LSB (bit 0) of 10 bit value
- 7 H Select frequency for det 1

$\textbf{Reg 9} \quad \textbf{IIC} \rightarrow \textbf{reg}$

bit

- 0 H Select frequency for det 1, Note : bit 3 of 10 bit value
- 1 H Select frequency for det 1
- 2 H Select frequency for det 1
- 3 H Select frequency for det 1
- 4 L Select frequency for det 1
- 5 H Select frequency for det 1
- 6 L Select frequency for det 1
- 7 H Select frequency for det 1 bit 9, MSB (10th bit) of 10 bit value



Reg 10 (Reg 0Ahex) $IIC \rightarrow reg$

bit

- 0 L Select data direction for S1 (L = input state)
- 1 L Select data direction for S2 (L = input state)
- 2 L Select data direction for S3 (L = input state)
- 3 L Text mux contril for LD1 (lock detect channel 1) test only!
- 4 L Text mux contril for LD2 (lock detect channel 1) test only!
- 5 Not used
- 6 L Select frequency for det 2, LSB (bit 0) of 10 bit value
- 7 L Select frequency for det 2

Reg 11 (Reg 0Bhex) IIC \rightarrow reg

bit

- 0 L Select frequency for det 2, Note : bit 3 of 10-bit value
- 1 L Select frequency for det 2
- 2 H Select frequency for det 2
- 3 L Select frequency for det 2
- 4 H Select frequency for det 2
- 5 H Select frequency for det 2
- 6 L Select frequency for det 2
- 7 H Select frequency for det bit 9, MSB (10th bit) of 10 bit value

Reg 12 (Reg 0Chex) IIC \rightarrow reg (read/write), Note : bits 6, 7 are Read only

bit

- 0 L Select 20kHz or REG12<4> (Pin 10 i/02)
- 1 L Not used
- 2 L Not used
- 3 L Select/Status of I/O 1
- 4 L Select/Status of I/O 2
- 5 L Not used
- 6 Read frequency of watchdog 1, LSB (bit 0) of 10 bit value
- 7 Read frequency of watchdog 1

Reg 13 (Reg 0Dhex) $\text{Reg} \rightarrow \text{IIC}$ (The watchdog is the vco frequency monitor)

- bit Read frequency of watchdog 1, Note : bit 3 of 10 bit value.
- 0 Read frequency of watchdog 1
- 1 Read frequency of watchdog 1
- 2 Read frequency of watchdog 1
- 3 Read frequency of watchdog 1
- 4 Read frequency of watchdog 1
- 5 Read frequency of watchdog 1
- 6 Read frequency of watchdog 1
- 7 Read frequency of watchdog 1, MSB (10th bit) of 10 bits

Reg 14 (Reg 0Ehex) IIC \rightarrow reg (read/write,) Note : bits 6, 7 are Read only

bit

- 0 L Select/Status of S1
- 1 L Select/Status of S2
- 2 L Select/Status of S3
- 3 Not used
- 4 Not used
- 5 Not used
- 6 Read frequ of Watchdog 2, Note : bit 0 of 10 bit value
- 7 Read frequ of Watchdog 2



Reg 15 (Reg 0Fhex) Reg \rightarrow **IIC** (The watchdog is the vco frequency monitor)

- Read frequency of watchdog 2, Note : bit 3 of 10 bit value bit
- Read frequency of watchdog 2 0
- 1 Read frequency of watchdog 2
- 2 Read frequency of watchdog 2
- 3 Read frequency of watchdog 2
- 4 Read frequency of watchdog 2
- 5 Read frequency of watchdog 2
- 6 Read frequency of watchdog 2
- 7 Read frequency of watchdog 2, MSB (10th bit) of 10 bits

Reg 16 (Reg 10hex) IIC \rightarrow reg

```
bit
```

- 0 L Select 5 bit audio volume control 00H = mute
- L Select 5 bit audio volume control 1 01H = -32.75dB2

 - Select 5 bit audio volume control L
- 3 Select 5 bit audio volume control L
- Select 5 bit audio volume control 4 L
- 5 L Not used
- 6 L Select 4.000MHz or 8.000MHz clock speed (L = 8MHz)
- 7 Select ANRS defeat (L = no ANRS, H = ANRS) 1

Reg 17 (Reg 11hex) IIC \rightarrow reg

bit

- 0 L Select LNB voltage. 8 bit value. LSB = bit 0 1 L Select LNB voltage. 8 bit value
- 2 Select LNB voltage. 8 bit value L
- 3 Select LNB voltage. 8 bit value L
- 4 Select LNB voltage. 8 bit value L
- 5 Select LNB voltage. 8 bit value L
- 6 L Select LNB voltage. 8 bit value
- Select LNB voltage. 8 bit value 7 L

Reg 18 (Reg 12hex) IIC \rightarrow reg (read/write)

- bit Not used
- 0 L Select video gain bits 00H = 0dB
- 1 L Select video gain bits 01H = +0.202dB
- 2 L Select video gain bits 02H = +0.404dB
- 3 L Select video gain bits n = +0.202 dB x n
- 4 L Select video gain bits 3FH = +12.73dB
- 5 L Select video gain bits
- L Sync enable (for noisy video Pin 52 only H = enabled) 6
- Not used 7

Reg 19 (Reg 13hex) IIC \rightarrow reg (read/write)

bit

- 0 L Select PLL loop gain bits, default value for 50kHz modulation
- 1 н Select PLL loop gain bits
- 2 Н Select PLL loop gain bits
- 3 L Select PLL loop gain bits
- Select PLL loop gain bits 4 н
- 5 L Not used
- 6 L Not used
- 7 Not used 1



1.25dB steps up to 1FH = +6db

00H = 0 volts ref

256 linear steps

FFH = 2.4 volts ref

Audio Mux Table

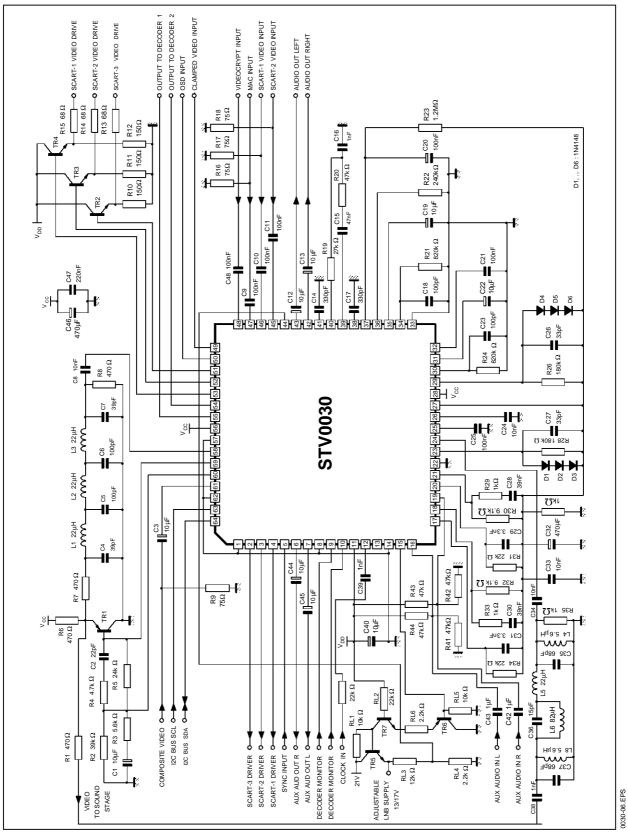
Register	2	2	į	5	6	
Bits	4 3 6 5		5	7	Function Selected	
	0 0 >		Х	Х	Х	mono left / channel 1
	1	0	Х	Х	Х	mono right / channel 2
	1 1 X X		Х	Х	stereo left & right	
	Х	Х	0	0	0	PLL audio 50µs de-emphasis
	Х	Х	Х	0	1	Aux audio - no de-amphasis
	Х	Х	0	1	Х	PLL audio J17 de-emphasis
	Х	Х	1	0	0	PLL audio 75µs de-emphasis
	X X 1 1		Х	PLL audio no de-emphasis		

Register 19 Truth Table for Programmable PLL Gain

	-	_		-	
4	3	2	1	0	Selected Nominal Carrier Modulation (Example 49.1kHz means ±49.1kHz)
0	0	0	0	0	cal : do not use = 0.3373V offset on VCO
0	0	0	0	1	cal : do not use = 0.3053V offset on VCO
0	0	0	1	0	cal : do not use = 0.2763V offset on VCO
0	0	0	1	1	calibration setting = 0.25V offset on VCO
0	0	1	0	0	296kHz modulation
0	0	1	0	1	267kHz modulation
0	0	1	1	0	242kHz
0	0	1	1	1	218kHz
0	1	0	0	0	198kHz
0	1	0	0	1	179kHz
0	1	0	1	0	161kHz
0	1	0	1	1	146kHz
0	1	1	0	0	133kHz
0	1	1	0	1	120kHz
0	1	1	1	0	109kHz
0	1	1	1	1	98.3kHz
1	0	0	0	0	89.7kHz
1	0	0	0	1	80.9kHz
1	0	0	1	0	73.1kHz
1	0	0	1	1	66.0kHz
1	0	1	0	0	60.0kHz
1	0	1	0	1	54.4kHz
1	0	1	1	0	49.1kHz = default power-up state
1	0	1	1	1	44.3kHz
1	1	0	0	0	39.8kHz
1	1	0	0	1	35.9kHz
1	1	0	1	0	32.4kHz
1	1	0	1	1	29.1kHz
1	1	1	0	0	26.7kHz
1	1	1	0	1	24.3kHz
1	1	1	1	0	21.9kHz
1	1	1	1	1	19.7kHz
L					1





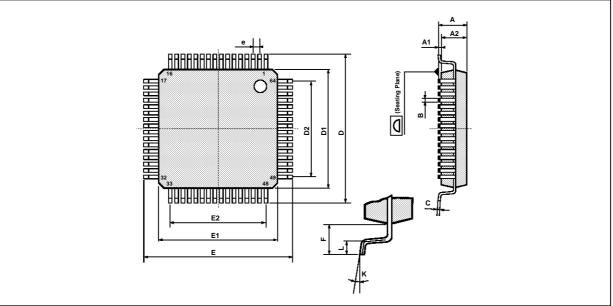


SGS-THOMSON

57.

PACKAGE MECHANICAL DATA

64 PINS - PLASTIC QUAD FLAT PACK



PQFP64.TBL

Dimensions		Millimeters				
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.
A			3.40			0.134
A1	0.25			0.01		
A2	2.55	2.80	3.05	0.10	0.11	0.12
В	0.30		0.45	0.012		0.018
С	0.13		0.23	0.005		0.009
D	16.95	17.20	17.45	0.667	0.677	0.687
D1	13.90	14.00	14.10	0.547	0.551	0.555
D2		12.00			0.472	
е		0.80			0.031	
E	16.95	17.20	17.45	0.667	0.677	0.687
E1	13.90	14.00	14.10	0.547	0.551	0.555
E2		12.00			0.472	
F		1.60			0.063	
К			0 ^o (min.),	7° (max.)		
L	0.65	0.80	0.95	0.025	0.031	0.037

Information furnished is believed to be accurate and reliable. However, SGS-THOMSON Microelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No licence is granted by implication or otherwise under any patent or patent rights of SGS-THOMSON Microelectronics. Specifications mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. SGS-THOMSON Microelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of SGS-THOMSON Microelectronics.

© 1994 SGS-THOMSON Microelectronics - All Rights Reserved

Purchase of I²C Components of SGS-THOMSON Microelectronics, conveys a license under the Philips I²C Patent. Rights to use these components in a I²C system, is granted provided that the system conforms to the I²C Standard Specifications as defined by Philips.

SGS-THOMSON Microelectronics GROUP OF COMPANIES

Australia - Brazil - China - France - Germany - Hong Kong - Italy - Japan - Korea - Malaysia - Malta - Morocco The Netherlands - Singapore - Spain - Sweden - Switzerland - Taiwan - Thailand - United Kingdom - U.S.A.

